

1.2A, HV Input Regulator with Output Current Limit for Battery Charger Applications

FEATURES

- **Wide Input Range:**
 - Operation from 7.5V to 30V
 - Overvoltage Lockout Protects Circuit Through 60V Transients
- **Operates with LTC Bat-Track™ Li-Ion Battery Chargers to Maximize Efficiency**
- **No Reverse Discharge Path from V_{OUT} to V_{IN}**
- **Programmable Output Current Limit**
- Integrated BOOST Diode
- Thermal Protection
- Small Application Size
- Thermally Enhanced 2mm × 3mm DFN Package

APPLICATIONS

- Battery Powered Equipment
- Portable USB Devices – Cameras, MP3 Players, PDAs
- Automotive Battery Regulation
- Multiple-Source Battery Chargers

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DESCRIPTION

The LT®3653 is a monolithic current mode PWM step-down regulator with programmable output current limit. Typically, it is used with battery charger power path controllers to pre-regulate the output and optimize the battery charger efficiency. Output current limit provides accurate control over system power dissipation.

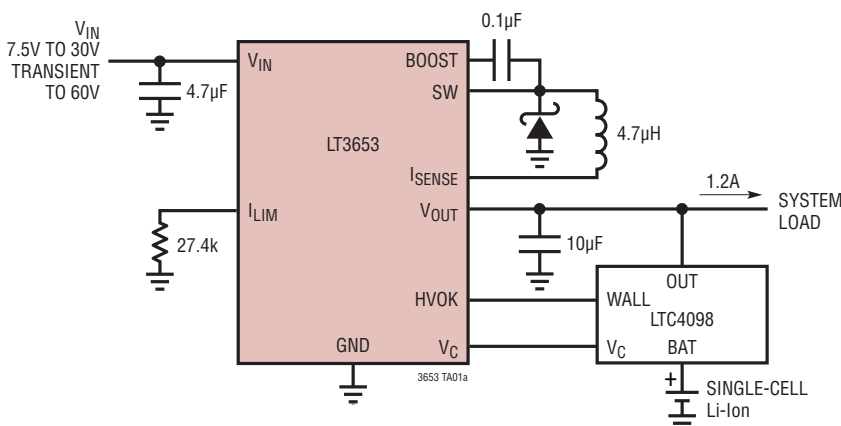
The wide operating input voltage range of 7.5V to 30V suits the LT3653 to a variety of input sources, including unregulated 12V wall adapters, 24V industrial supplies, Firewire and automotive power. Input overvoltage protection allows the LT3653 to handle 60V input transients.

The LT3653 control pin, V_C , is available for external battery tracking control, Bat-Track. The LT3653 will block output discharge during an open or shorted input.

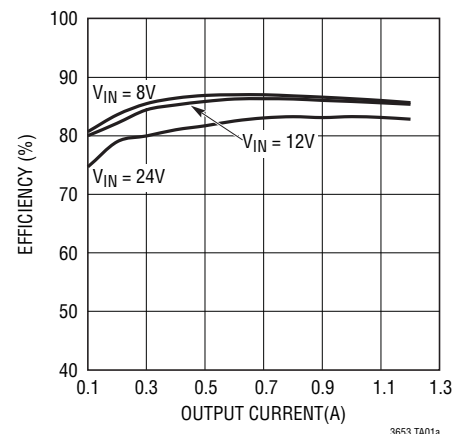
The HVOK pin indicates that the internal bias supplies are present and no faults have occurred (i.e., overtemperature and input overvoltage and undervoltage). The LT3653 includes a 1.5MHz frequency oscillator, internal compensation, and an internal boost diode to minimize the number of external components. The LT3653 is available in an 8-lead (2mm × 3mm) package with exposed pads for low thermal resistance.

TYPICAL APPLICATION

High Voltage Single-Cell Battery Charger Application



Efficiency vs Load Current



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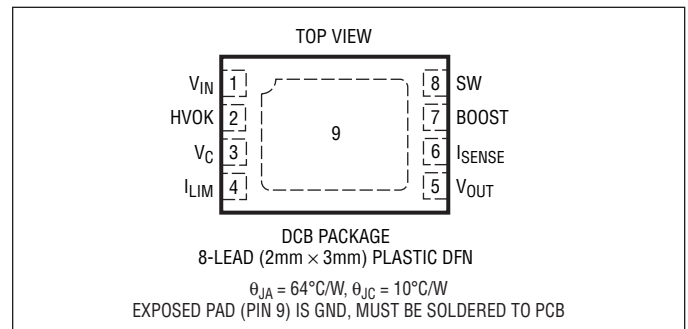
LT3653

ABSOLUTE MAXIMUM RATINGS

(Note 1)

Input Voltage (V_{IN}) (Note 3)	60V
Boost Pin Voltage (BOOST)	50V
Boost Pin Above SW Pin	25V
HVOK, V_{OUT} , I_{SENSE} Pins	6V
V_C Pin	3V
Operating Junction Temperature Range (Notes 2, 5)	
LT3653E	-40°C to 125°C
LT3653I	-40°C to 125°C
Storage Temperature	-65°C to 150°C

PIN CONFIGURATION



ORDER INFORMATION

LEAD FREE FINISH	TAPE AND REEL	PART MARKING*	PACKAGE DESCRIPTION	TEMPERATURE RANGE
LT3653EDCB#PBF	LT3653EDCB#TRPBF	LDJN	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 125°C
LT3653IDCB#PBF	LT3653IDCB#TRPBF	LDJN	8-Lead (2mm × 3mm) Plastic DFN	-40°C to 125°C

Consult LTC Marketing for parts specified with wider operating temperature ranges. *The temperature grade is identified by a label on the shipping container. Consult LTC Marketing for information on non-standard lead based finish parts.

For more information on lead free part marking, go to: <http://www.linear.com/leadfree/>

For more information on tape and reel specifications, go to: <http://www.linear.com/tapeandreel/>

ELECTRICAL CHARACTERISTICS

The ● denotes the specifications which apply over the full operating temperature range, otherwise specifications are at $T_A = 25^{\circ}\text{C}$; $V_{IN} = 12\text{V}$, $V_{BOOST} = 17\text{V}$, $V_{SW} = 0\text{V}$, $R_{ILIM} = 33.2\text{k}$, unless otherwise noted.

PARAMETER	CONDITIONS	MIN	TYP	MAX	UNITS
V_{IN} Operating Range	●	7.5		30	V
V_{IN} Undervoltage Lockout	Rising	6.5	7	7.5	V
V_{IN} Undervoltage Hysteresis			500		mV
V_{IN} Overvoltage Lockout	Rising	30	33	36	V
V_{IN} Overvoltage Lockout Hysteresis			1		V
V_{IN} Supply Current	Not Switching		2.8	3.4	mA
V_{OUT} Reverse Leakage Current	$V_{IN} = 0\text{V}$, $V_{OUT} = 4.8\text{V}$ $V_{IN} = \text{Open}$, $V_{OUT} = 4.8\text{V}$			1 1	μA μA
V_{OUT} Nominal Set Point	V_C Pin = Open		4.8	5	V
V_{OUT} Current Limit Range	●	0.4		1.2	A
V_{OUT} Current Limit	$R_{ILIM} = 33.2\text{k}$	0.85 0.80	1 1	1.15 1.2	A A
Operating Frequency	●	1350	1500	1650	kHz
V_C Source Current	$V_{OUT} = 0\text{V}$, $V_C = 1.5\text{V}$		12		μA
V_C Sink Current	$V_{OUT} = 5\text{V}$, $V_C = 1.5\text{V}$		10		μA
V_C Switching Threshold	$I_{OUT} = 0\text{mA}$		1.1		V

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PARAMETER	CONDITIONS		MIN	TYP	MAX	UNITS
V_C Clamp Voltage	$V_{OUT} = 0\text{V}$			1.75		V
Switch Peak Current Limit	(Note 4)			2		A
Switch V_{CESAT}	$I_{SW} = 1\text{A}$			275		mV
Boost Diode Drop	$I_{DIODE} = 60\text{mA}$			0.9	1.2	V
Boost Pin Current	$I_{SW} = 1\text{A}$			28		mA
HVOK Output Voltage High	$I_{HVOK} = 1\text{mA}$	●	4	4.4	5	V
HVOK Output Voltage Low	$I_{HVOK} = -2\mu\text{A}$	●			0.25	V

Note 1: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. Exposure to any Absolute Maximum Rating condition for extended periods may affect device reliability and lifetime.

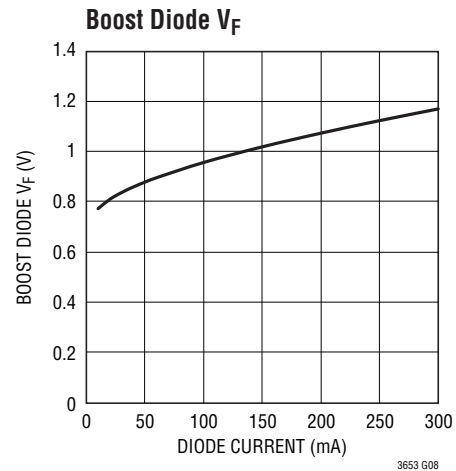
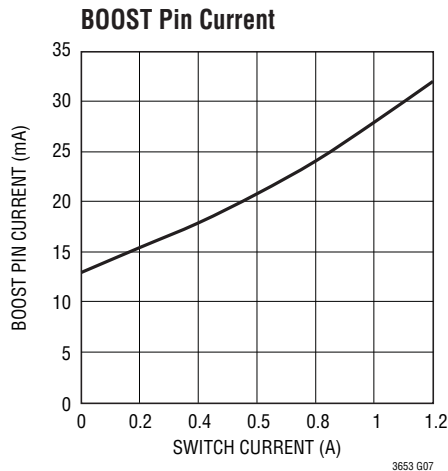
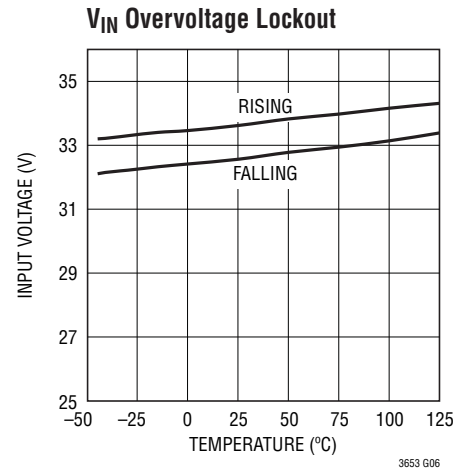
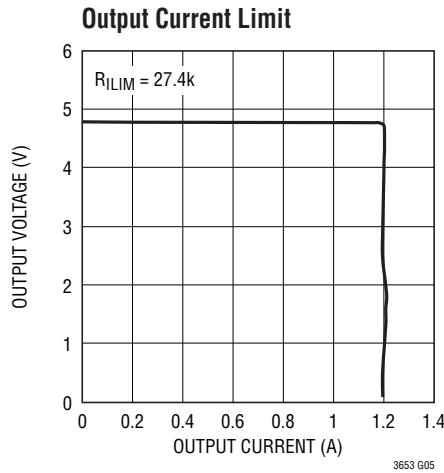
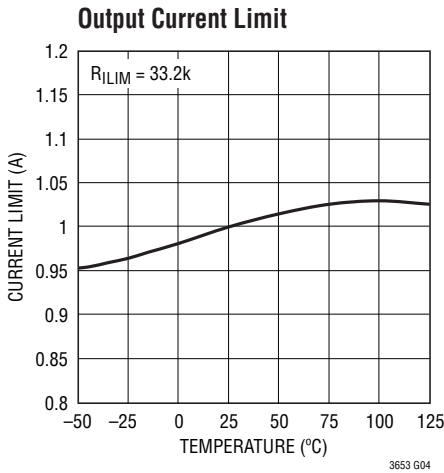
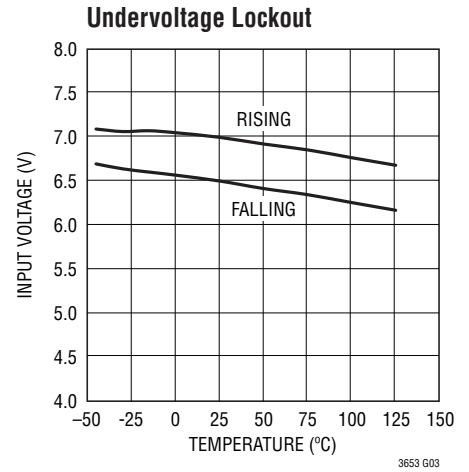
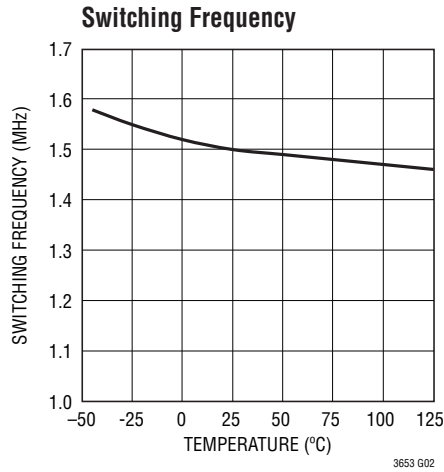
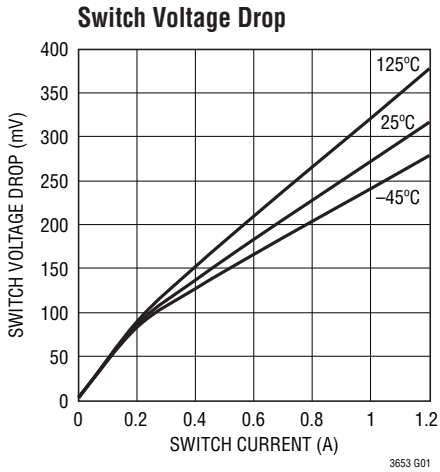
Note 2: The LT3653E is guaranteed to meet performance specifications from 0°C to 125°C . Specifications over the -40°C to 125°C operating temperature range are assured by design, characterization, and correlation with statistical process controls. The LT3653I specifications are guaranteed over the full -40°C to 125°C temperature range. High junction temperatures degrade operating lifetimes. Operating lifetime is derated at junction temperatures greater than 125°C .

Note 3: Absolute Maximum Voltage at V_{IN} pin is 60V for non-repetitive 1 second transients.

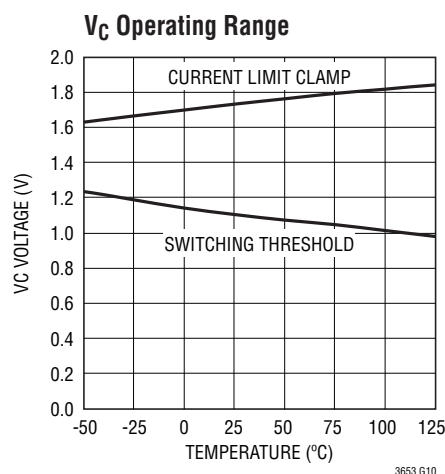
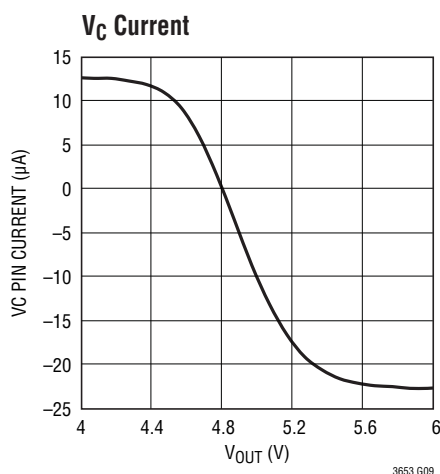
Note 4: Switch Peak Current Limit guaranteed by design and/or correlation to static test.

Note 5: This IC includes overtemperature protection that is intended to protect the device during momentary overload conditions. Junction temperature will exceed the maximum operating junction temperature when overtemperature protection is active. Continuous operation above the specified maximum operating junction temperature may impair device reliability.

TYPICAL PERFORMANCE CHARACTERISTICS T_A = 25°C unless otherwise noted



TYPICAL PERFORMANCE CHARACTERISTICS $T_A = 25^\circ\text{C}$ unless otherwise noted



PIN FUNCTIONS

V_{IN} (Pin 1): The V_{IN} pin supplies current to the LT3653's internal regulator and the internal power switch. Locally bypass this pin with a capacitor.

HVOK (Pin 2): The HVOK pin is a status pin which indicates that the internal bias rail is present and that an input undervoltage lockout fault, an overvoltage lockout fault or an over temperature fault are not present.

V_C (Pin 3): The V_C pin is the output of the internal error amplifier and is internally compensated. The voltage on this pin controls the peak switch current. Connect this pin to the battery charger V_C pin for battery tracking function.

I_{LIM} (Pin 4): Connect a resistor to GND to program the output current limit of the regulator. See the I_{LIM} Resistor section in Application Information.

V_{OUT} (Pin 5): The V_{OUT} pin is connected to the negative terminal of the internal current sense resistor of the output current limit circuit. Maximum V_{OUT} voltage is regulated to 4.8V.

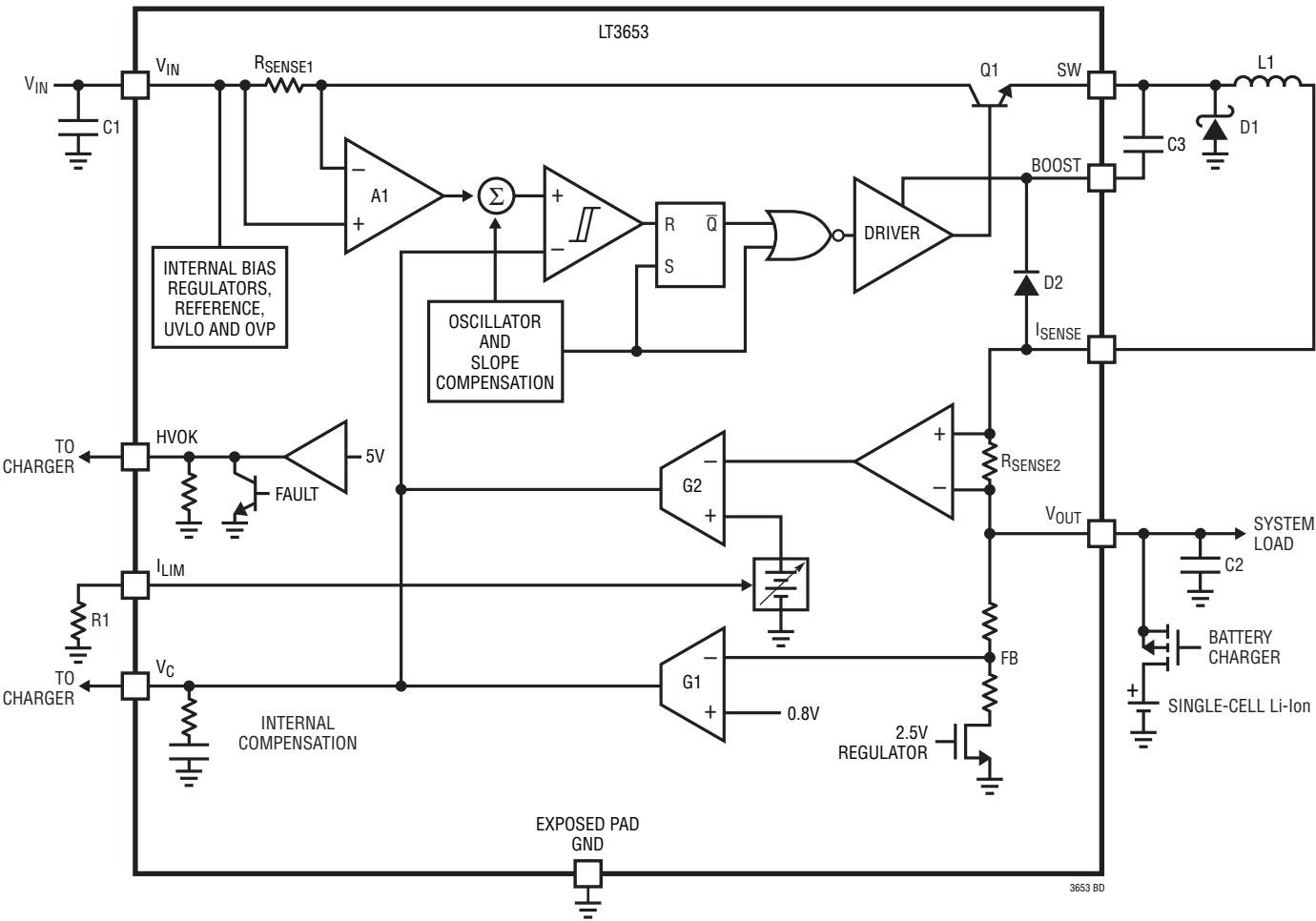
I_{SENSE} (Pin 6): The I_{SENSE} pin is the positive input to the internal current sense resistor of the output current limit circuit. The I_{SENSE} pin is also the anode of the internal BOOST diode.

BOOST (Pin 7): The BOOST pin provides a drive voltage, higher than the input voltage, to the internal power switch. The BOOST pin is internally connected to the cathode of the BOOST diode.

SW (Pin 8): The SW pin is the output of the internal power switch. Connect this pin to the inductor, catch diode and boost capacitor.

Exposed Pad (Pin 9): Ground. The Exposed Pad must be soldered to the PCB and electrically connected to ground. Use a large ground plane and vias to optimize thermal performance.

BLOCK DIAGRAM



3653 BD

OPERATION

Please refer to the Block Diagram. The LT3653 is specifically tailored for use with a battery charger power path controller. The LT3653 regulates the high voltage input to a lower voltage to provide power to the system load and to the single-cell Li-Ion battery charger. High efficiency in the battery charger is achieved by keeping the voltage across it low. To accomplish this, the battery charger takes control of the LT3653 regulator control node, V_C , and overrides the error amp, G1. The output voltage is regulated by the battery charger to a voltage slightly above the battery, typically 300mV.

HVOK is a status pin which indicates to the charger that a high voltage input is present and that the LT3653 is ready to start providing power to the system load. When the HVOK pin is low, the LT3653 is not switching and the system output is not supported by the LT3653 regulator.

The LT3653 is a constant frequency, current mode step down regulator. A switch cycle is initiated when the 1.5MHz oscillator enables the RS flip flop, turning on the internal power switch, Q1. The sense amplifier (A1) monitors the switch current via the voltage dropped across the current sense resistor R_{SENSE1} . The comparator compares the amplified current signal with the output (V_C) of the error amplifier (G1). The switch is turned off when this current exceeds a value determined by the V_C voltage. The error amplifier monitors the V_{OUT} voltage through an internal resistor divider and, when not driven externally servos the V_C voltage to regulate V_{OUT} . If the V_{OUT} voltage drops, the V_C voltage will be driven higher increasing the output current and V_{OUT} voltage.

An active clamp (not shown) on the V_C node provides current limit. The LT3653 is internally compensated with a pole zero combination.

An external capacitor and internal diode, D2, are used to generate a voltage at the BOOST pin that is higher than the input supply. This allows the driver to fully saturate the internal bipolar NPN power switch for efficient operation. The switch driver operates from either V_{IN} or BOOST to ensure startup.

An internal regulator provides power to the control circuitry. This regulator includes input undervoltage and overvoltage protection which disables switching action when V_{IN} is less than 7V and greater than 33V, typical. When switching is disabled, the LT3653 safely sustains input voltages up to 60V. Note that while switching is disabled the output will discharge.

Output current limiting is provided via the servo action of amplifier G2. The voltage across the sense resistor, R_{SENSE2} , is compared to a voltage programmed by external resistor R1 on the I_{LIM} pin. A capacitor averages the inductor ripple current. If the averaged inductor current exceeds the programmed value then the V_C voltage is pulled low, reducing the current in the regulator. The output current limit circuit allows for lower current rated power path components and provides better control of system power dissipation.

APPLICATIONS INFORMATION

Inductor Selection

A 4.7μH inductor is recommended for most LT3653 applications. This value provides a good tradeoff between size and ripple current. The inductor's RMS current rating must be greater than the maximum load current and its saturation current should be about 30% higher. The output current limit circuit tightly controls the maximum average inductor current therefore the inductor RMS current rating does not have to be overrated to handle short-circuit or overload conditions. For high efficiency, keep the series resistance (DCR) less than 0.1Ω. Output voltage ripple can be reduced by using a higher value inductor. The cost is a larger physical size and poorer transient response. A lower value inductor has higher ripple currents but is physically smaller or, for the same size, has lower DCR usually resulting in higher efficiency.

Catch Diode

The catch diode conducts current only during switch off time. Average forward current in normal operation is calculated from:

$$I_{D(AVG)} = I_{OUT} \frac{V_{IN} - V_{OUT}}{V_{IN}}$$

where I_{OUT} is the maximum output load current programmed by the I_{LIM} resistor. Peak reverse voltage is equal to the regulator input voltage. Use a Schottky diode with a reverse voltage rating greater than the maximum input voltage. The overvoltage protection feature in the LT3653 keeps the switch off when $V_{IN} > 33V$ (typical), allowing the use of a 40V rated Schottky, even when V_{IN} ranges up to 60V.

Input Capacitor

Bypass the input of the LT3653 circuit with a 4.7μF or higher value ceramic capacitor of X7R or X5R type. Y5V types have poor performance over temperature and applied voltage and should not be used. If the input power source has high impedance, or there is significant inductance due to long wires or cables, additional bulk capacitance may be necessary. This can be provided with a low performance electrolytic capacitor. Step-down regulators draw current from the input supply in pulses with very fast rise and fall times.

The input capacitor is required to reduce the resulting voltage ripple at the LT3653 and to force this very high frequency switching current into a tight local loop, minimizing EMI. Place the capacitor in close proximity to the LT3653 and the catch diode (see the PCB Layout section).

Output Capacitor

A 10μF or greater ceramic capacitor is required for low output ripple and good transient response. Ceramic capacitors have very low equivalent series resistance (ESR) and provide the best ripple performance. Use X5R or X7R types and keep in mind that a ceramic capacitor biased with V_{OUT} has less than its nominal capacitance.

High performance electrolytic capacitors can be used for the output capacitor. Low ESR is important, so choose one that is intended for use in switching regulators. Keep the ESR less than 0.1Ω.

I_{LIM} Resistor

The LT3653 output current limit controls the maximum current delivered from the LT3653 regulator. This allows tighter control of the system power dissipation and also protects the inductor and diode from overheating during an overload or short-circuit condition. A resistor connected from the I_{LIM} pin to GND programs the output current limit. Table 1 details the I_{LIM} resistor values for specific desired output current limits

Table 1. Output Current Limit vs R_{ILIM} Value

OUTPUT CURRENT LIMIT (A)	R_{ILIM} VALUE (kΩ)
0.4	121
0.6	66.5
0.8	45.3
1	33.2
1.2	27.4

Boost Capacitor Selection

The boost capacitor is calculated with the following formula:

$$C_{BOOST} = \frac{(I_{OUT(MAX)}) \cdot V_{OUT}}{0.1V \cdot 30 \cdot V_{IN} \cdot 1.5MHz}$$

Typically, a 0.1μF capacitor is used.

APPLICATIONS INFORMATION

Battery Charger Operation

Connect the control node, V_C pin, of the LT3653 to the V_C pin of the battery charger power path controller. The V_C node is internally clamped; however, take care not to overdrive the pin. The LT3653 is internally compensated with a pole zero combination on the output of the g_m amplifier, G1. Check stability over the full input voltage range, output load range and temperature.

Connect the HVOK node of the LT3653 to the high voltage present pin of the charger. This is the WALL pin on the LTC4098. The HVOK pin is capable of supplying up to 1mA of drive current. When the HVOK pin is low the LT3653 is not switching and the system output cannot be supported by the LT3653 regulator. See the Typical Applications section for different configurations.

PCB Layout

Proper operation and minimum EMI requires a careful printed circuit board layout. Figure 1 shows the recommended component placement with trace, ground plane and via locations. Note that large, switched currents flow in the LT3653's V_{IN} and SW pins, the catch diode (D1) and the input capacitor (C1). Keep the loop formed by these components as small as possible and tied to system ground in only one place. Place these components, along with the inductor and output capacitor, on the same side of the circuit board, with their connections made on that layer. Place a local, unbroken ground plane below these

components, and tie this ground plane to system ground at one location, ideally at the ground terminal of the output capacitor C2. Make the SW and BOOST nodes as short as possible. Include vias near the exposed GND pad of the LT3653 to help remove heat from the LT3653 to the ground plane.

High Temperature Considerations

The die temperature of the LT3653 must not exceed the maximum rating of 125°C. This is generally not a concern unless the ambient temperature is above 85°C. For higher temperatures, take care in the layout of the circuit to ensure good heat sinking of the LT3653. Derate the maximum load current as the ambient temperature approaches 125°C. The die temperature is calculated by multiplying the LT3653 power dissipation by the thermal resistance from junction to ambient. Power dissipation within the LT3653 is estimated by calculating the total power loss from an efficiency measurement and subtracting the catch diode loss. Thermal resistance depends on the layout of the circuit board, but 64°C/W is typical for the (2mm × 3mm) DFN (DCB) package.

Other Linear Technology Publications

Application Notes 19, 35 and 44 contain more detailed descriptions and design information for Buck regulators and other switching regulators. The LT1376 data sheet has a more extensive discussion of output ripple, loop compensation and stability testing.

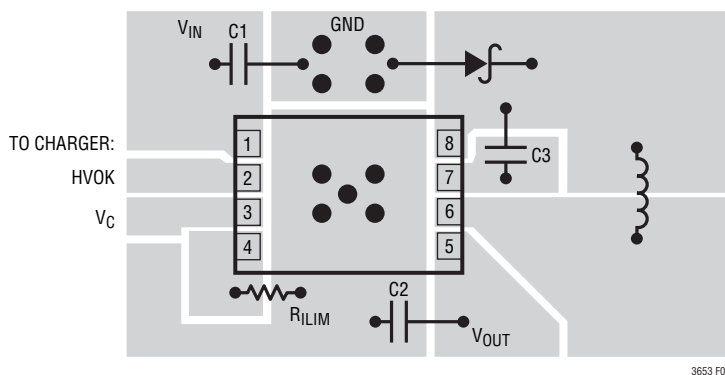
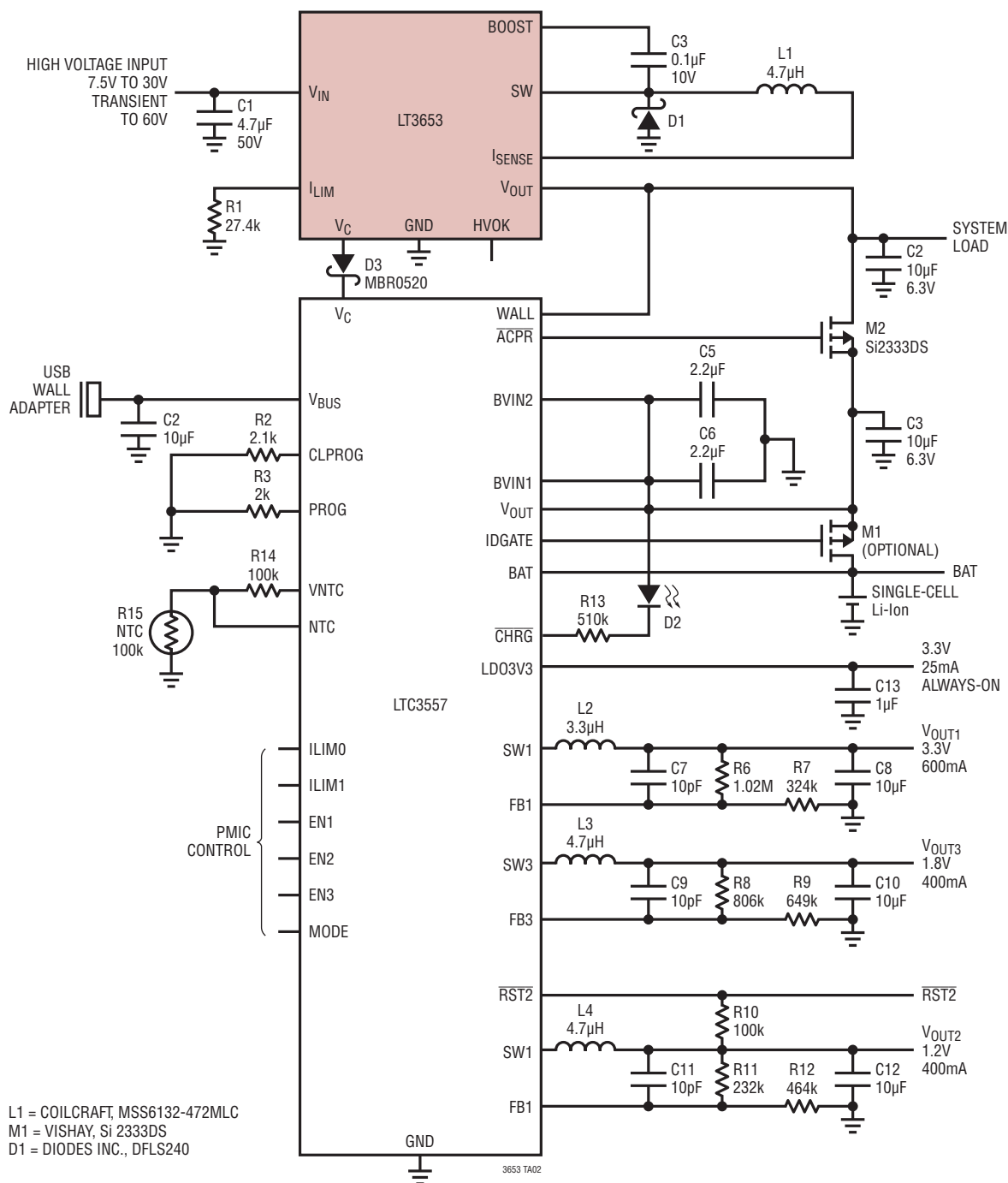


Figure 1. LT3653 PCB Layout

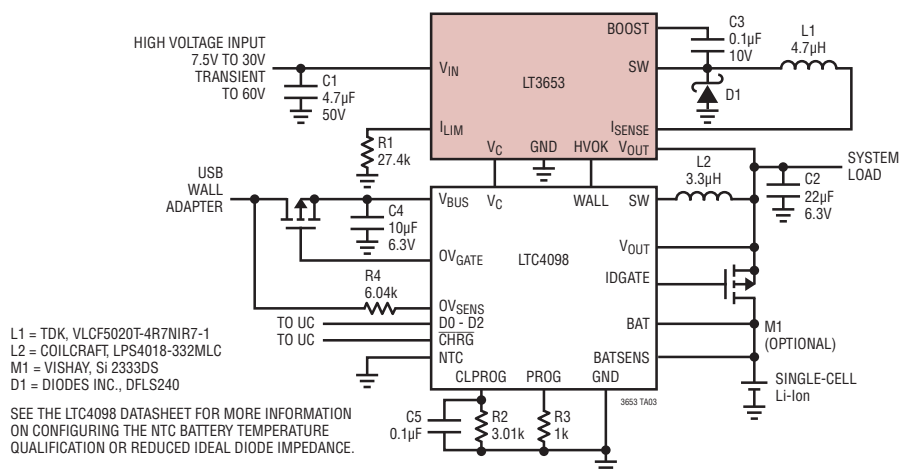
TYPICAL APPLICATION

Dual Input High Efficiency Battery Charger and Triple Buck PMIC Application



TYPICAL APPLICATION

Dual Input High Efficiency Battery Charger with Low Battery Start-Up and 1.2A Output Current Limit



RELATED PARTS

PART NUMBER	DESCRIPTION	COMMENTS
LT1939	Dual 25V, 2A (I_{OUT}), 2.5MHz, High Efficiency DC/DC Converter and LDO Controller	V_{IN} : 3.6V to 25V, $V_{OUT(MIN)}$ = 0.8V, I_Q = 2.5mA, I_{SD} < 10µA, 3mm × 3mm DFN-10 Package
LT3480	36V with Transient Protection to 60V, 2A (I_{OUT}), 2.4MHz, High Efficiency Step-Down DC/DC Converter with Burst Mode Operation	V_{IN} : 3.6V to 38V, $V_{OUT(MIN)}$ = 0.78V, I_Q = 70µA, I_{SD} < 1µA, 3mm × 3mm DFN-10 and TSSOP-16E Packages
LT3500	36V, 40V _{MAX} , 2A, 2.5MHz High Efficiency DC/DC Converter and LDO Controller	V_{IN} : 3.6V to 36V, $V_{OUT(MIN)}$ = 0.8V, I_Q = 3.7mA, I_{SD} < 10µA, TSSOP-20E Package
LT3505	36V with Transient Protection to 40V, 1.4A (I_{OUT}), 3MHz, High Efficiency Step-Down DC/DC Converter	V_{IN} : 3.6V to 34V, $V_{OUT(MIN)}$ = 0.78V, I_Q = 2mA, I_{SD} < 2µA, 3mm × 3mm DFN-8 and MSOP-8E Packages
LT3507	36V, Triple 2.5MHz, 2.4A + 1.5A + 1.5A (I_{OUT}) and LDO Controller	V_{IN} : 4V to 36V, $V_{OUT(MIN)}$ = 0.8V, I_Q = 7mA, I_{SD} < 1µA, 5mm × 7mm QFN-38 Package
LT3508	36V with Transient Protection to 40V, Dual 1.4A (I_{OUT}), 3MHz, High Efficiency Step-Down DC/DC Converter	V_{IN} : 3.7V to 37V, $V_{OUT(MIN)}$ = 0.8V, I_Q = 4.6mA, I_{SD} < 1µA, 4mm × 4mm QFN-24 and TSSOP-16E Packages
LT3557/-1	USB Power Manager with Li-Ion/Polymer Charger, Triple Synchronous Buck Converter + LDO	Complete Multi-Function PMIC: Linear Power Manager & Three Buck Regulators Bat-Track Adaptive Output Control, "-1" version has 4.1V Vfloat, 4mm x 4mm QFN-28 Package
LT3680	36V, 3.5A (I_{OUT}) 2.4MHz, High Efficiency Step-Down DC/DC Converter	V_{IN} : 3.6V to 36V, $V_{OUT(MIN)}$ = 0.8V, I_Q = 75mA, I_{SD} < 1mA, 3mm × 3mm DFN-10, MSOP-10E
LT3685	36V with Transient Protection to 60V, 2A (I_{OUT}), 2.4MHz, High Efficiency Step-Down DC/DC Converter	V_{IN} : 3.6V to 38V, $V_{OUT(MIN)}$ = 0.78V, I_Q = 70µA, I_{SD} < 1µA, 3mm × 3mm DFN-24 and MSOP-10E Packages
LT4089/-1/-5	High Voltage USB Power Manager with Ideal Diode Controller and High Efficiency Li-Ion Battery Charger	High Efficiency 1.2A Charger from 6V to 36V (40V max) Input; "-1" for 4.1V float voltage batteries; Bat-Track™ Adaptive Output Control (LTC4089); Fixed 5V Output (LTC4089-5, -1); 3mm × 6mm DFN-22 Package
LT4090/-5	High Voltage USB Power Manager with Ideal Diode Controller and High Efficiency Li-Ion Battery Charger	High Efficiency 1.2A Charger from 6V to 38V (60V max) Input Bat-Track Adaptive Output Control; "-5" has no Bat-Track; 3mm × 6mm DFN-22 Package
LTC4098	USB-Compatible Switchmode Power Manager with OVP	High V_{IN} : 38V operating, 60V transient; 66V OVP Maximizes Available Power from USB Port, Bat-Track, "Instant-ON" Operation, 3mm x 4mm UTQFN-20 Package

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